

DC-DC LED Constant Current Regulator



7W 250mA High Efficiency Step Down LED Driver

Power Your Idea

Features

- RoHS-compliant 14 Pin DIL Package (Row Dist :5.08mm)
- Constant Current Output (±8% Output Current Accuracy)
- LED Driver Current up to 250mA
- Power LED Driver
- Wide Input Voltage Range: 7V to 30V (40V for 0.5sec.)
- Output Power to 7W
- Driver LED Strings of up to 28V (2V to 28V)
- High Efficiency (up to 95%)
- PWM/Digital Dimming and Analog Voltage Dimming
- Open and Short LED Protection
- -40°C ~ 85°C Operation Temperature Range



Application

- 12V and 24V Lighting Systems
- Household/Commercial lighting
- Suitable for high illumination LED
- Power limited (battery) lighting system

GP-LD2528-0M is a high efficiency step-down converter optimized to drive high current LEDs. The control algorithm allows highly efficient and accurate LED current regulation. The device operates from an input 7Vdc to 30Vdc and provides an externally adjustable output current of up to 250mA and output power up to 7 watts. Compact size of DIL14 allows designer to integrate this driver together with LED module. UL94V-0 grade molded case with high grade filling material provide excellent fire proof characteristics.

Electrical Specifications	
Input Voltage (Vdc)	7V ~ 30V, 24Vdc Nominal
Input Filter	Capacitor
Input Current (No-Load)	2.0mA
Output Voltage Range (Vin = 30V)	2V to 28V
Output Current Range (Vin - Vout > 2V to 3V)	250mA
Output Current Accuracy (Iout = 250mA)	±8%
Output Power	7W,
Ripple and Noise, (20 MHz bandwidth)	200mVpk-pk, max.
Maximum Efficiency at Full Load	95%
Capacitive Load	47uF
Operating Frequency	70kHz ~ 490kHz
Short Circuit Protection	Regulated at Rated Output Current
Temperature Coefficient (Ta = -40°C to +85°C)	±0.03%/°C, max.
Thermal Impedance (Nature Convection)	+35°C/W
Safety Standard : (designed to meet)	IEC / EN 60950-1

Environmental Specifications	
Operating Temperature Range, Ta	-40°C to +85°C
Storage Temperature Range	-40°C to +125°C
Humidity	95% rel H
Maximum Case Temperature	+105°C
Cooling	Nature Convection
Reliability Calculated MTBF (MIL-HDBK-217F)	>950 Khrs
Soldering Temperature (1.5mm from case 10 sec. max.)	+260°C, max.

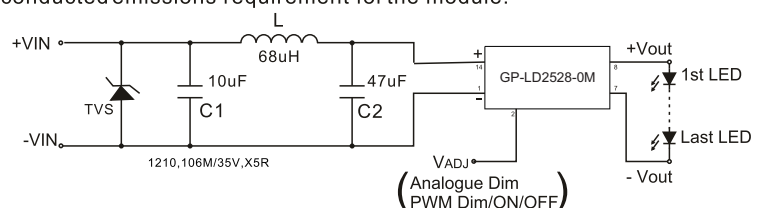
Dimming Control and ON/OFF Control (Leave Open if Not Used):	
V _{ADJ} Pin Input Voltage Range	0V to 1.25V
V _{ADJ} Pin Drive Current (V _{ADJ} = 1.25V)	<1mA
Analog Dimming	
Adjust Output Current (Vin - Vout < 20V)	25% to 100%
Control Voltage Range Limits	
On	0.3V < V _{ADJ} < 1.25V
Off	V _{ADJ} < 0.15V
PWM Dimming	
Recommended Maximum Operation Frequency	1KHz
Adjust Output Current	0% to 100%
Remote ON/OFF	
DC/DC ON	0.3V < V _{ADJ} < 1.25V or open circuit
DC/DC OFF (Shutdown)	V _{ADJ} < 0.15V or Short circuit pin 1 and pin 2
Quiescent Input Current in Shutdown Mode (Vin = 30V)	25µA, max.

EMC Specifications	
EMI Radiated & Conducted Emissions	EN 55015 (CISPR22)
EMS Immunity	EN 61547
IEC 61000-4-2	Perf. Criteria A
IEC 61000-4-3	Perf. Criteria A
IEC 61000-4-4	Perf. Criteria A
IEC 61000-4-6	Perf. Criteria A
IEC 61000-4-8	Perf. Criteria A

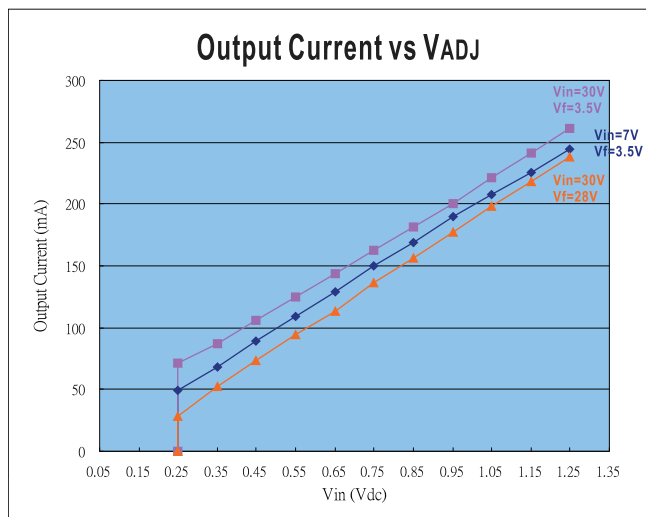
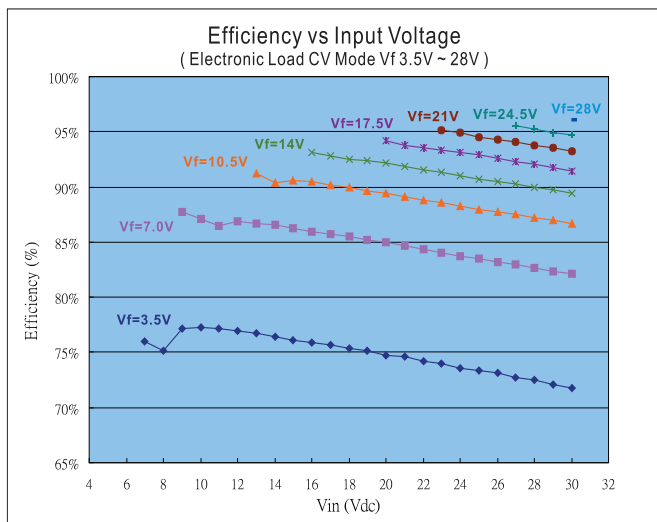
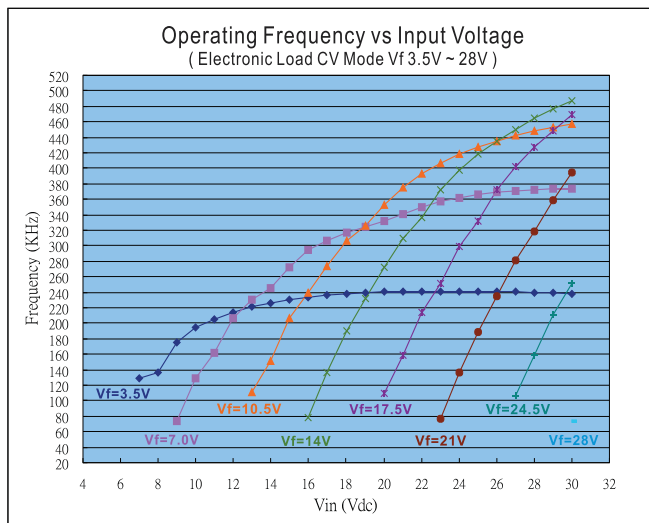
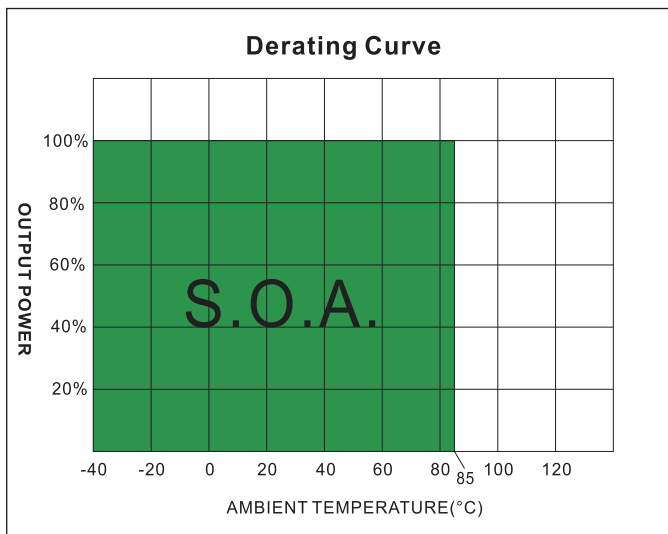
Physical Specifications	
Case Material	Non-Conductive Black Plastic (UL94V-0 rated)
Potting Material	Epoxy (UI94V-0 rated)
Pin Material	Ø0.5mm Brass Solder-coated
Weight	2.6g
Dimensions	0.80"x0.40"x0.27"

NOTE

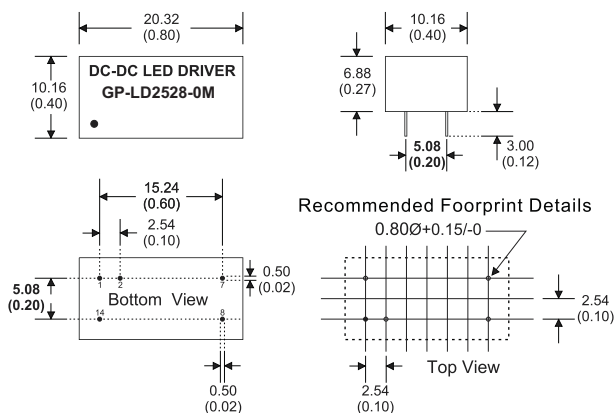
1. Reversed power source damages the circuit, No connection is allowed between input ground and output.
2. DO NOT operate the driver over 7W output.
3. Leave pin V_{ADJ} open if not in use, ground pin to shut down the converter. Connecting V_{adj} to V_{in} damages the circuit.
4. Maximum output open voltage is equal to input voltage.
5. Input filter components (C1, L, C2) are used to help meet conducted emissions requirement for the module.
6. For the compliance with IEC61000-4-5, a TVS is thus recommended to be installed in front of the input filter, the reference model: 3.0SMCJ24A or SMCJ24A (TVS Max Clamping Voltage @ Max Peak Pulse Current VC (V) ≤ 40V)



Typical Operating Conditions



MECHANICAL DIMENSION



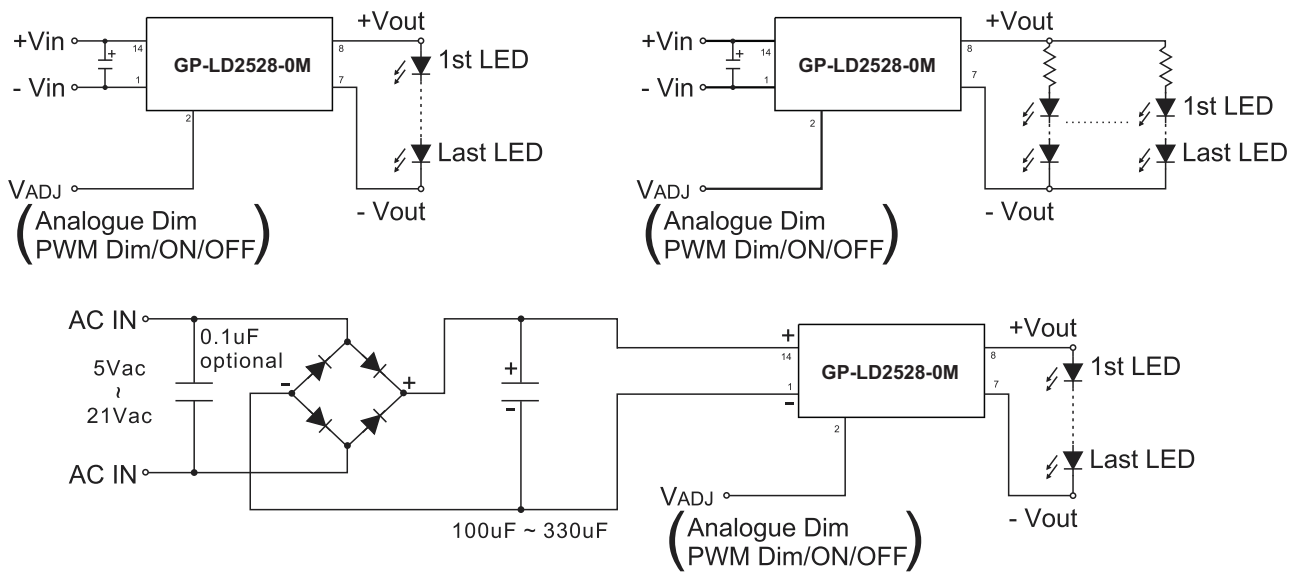
14 Pin DIL Package

- Notes : All dimensions are typical in millimeters (inches).
1. Pin diameter: 0.5 \pm 0.05 (0.02 \pm 0.002)
 2. Pin pitch and length tolerance: \pm 0.35 (\pm 0.014)
 3. Case Tolerance: \pm 0.5 (\pm 0.02)

Pin #	CONNECTIONS	
1	- V Input	- DC Supply
2	VADJ	PWM/ON/OFF or not used
7	- V Output	LED Cathode Connection
8	+V Output	LED Anode Connection
14	+V Input	+DC Supply

No connection is allowed between input and output

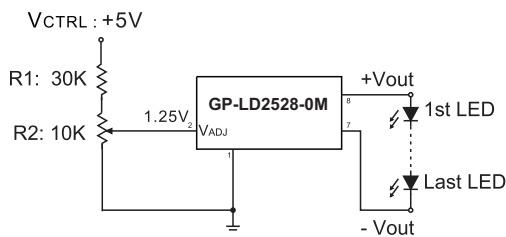
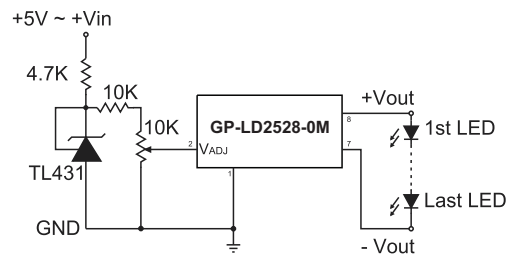
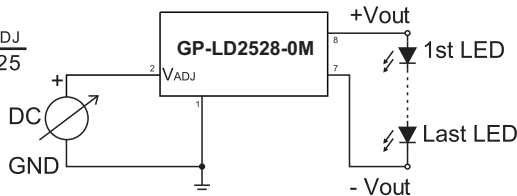
Typical application



Output Current Adjustment By External DC Control Voltage

The nominal output current is then given by:

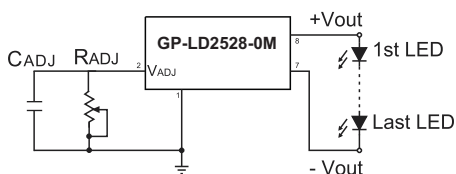
$$I_{outnom} \approx I_{out} \times \frac{V_{ADJ}}{1.25}$$



$$V_{ADJ} = \frac{R2}{R1 + R2} \times V_{CTRL}$$

Resistor dimming

By connecting a variable resistor between ADJ and GND, simple dimming can be achieved. Capacitor CADJ is optional for better AC mains interference and HF noise rejection. Recommend value of CADJ is 0.22uF.



The current output I_{outnom} can be determined using the equation:

$$I_{outnom} = \frac{I_{out} \times R_{ADJ}}{(R_{ADJ} + 200K)}$$

If the value of R_{ADJ} is 0 to 2M ohm, the maximum adjust range of output current is 25% to 90%. (For $V_{in}-V_{out} < 20V_{dc}$)

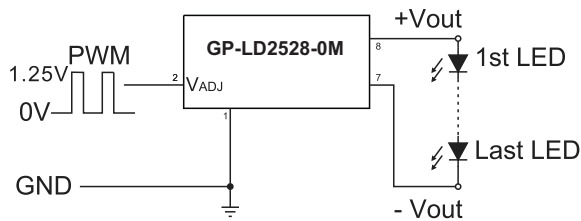
Typical application

Output Current Adjustment By PWM Control

Directly driving ADJ input

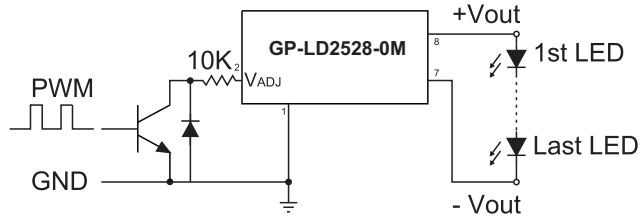
A Pulse Width Modulated (PWM) signal with duty cycle, DP_{PWM} , can be applied to the ADJ pin, as shown below

$$I_{out_{nom}} \approx I_{out} \times DP_{PWM} \quad [\text{If PWM frequency} < 200\text{Hz, for } 0.1 < DP_{PWM} < 1]$$



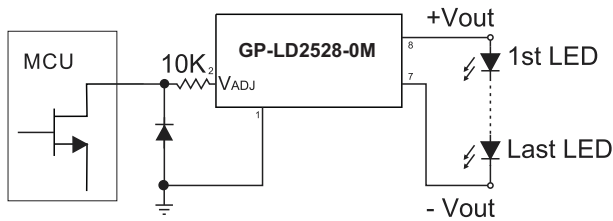
Driving the ADJ input via open collector transistor

The diode and resistor suppress possible high amplitude negative spikes on the ADJ input resulting from the drain-source capacitance of the transistor. Negative spikes at the input to the device should be avoided as they may cause errors in output current, or erratic device operation.



Driving the ADJ input from a microcontroller

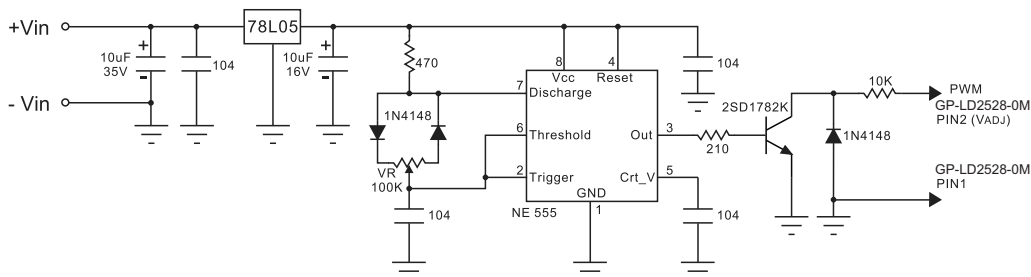
Another possibility is to drive the device from the open drain output of a microcontroller. The diagram below shows one method of doing this:



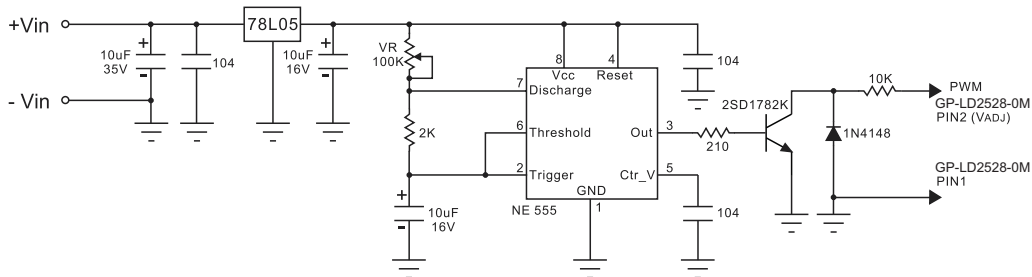
The diode and resistor suppress possible high amplitude negative spikes on the ADJ input resulting from the drain-source capacitance of the FET. Negative spikes at the input to the device should be avoided as they may cause errors in output current, or erratic device operation.

Output Current Adjustment By PWM Control (Dimming)

To avoid visible flicker the PWM signal must be greater than 100Hz.

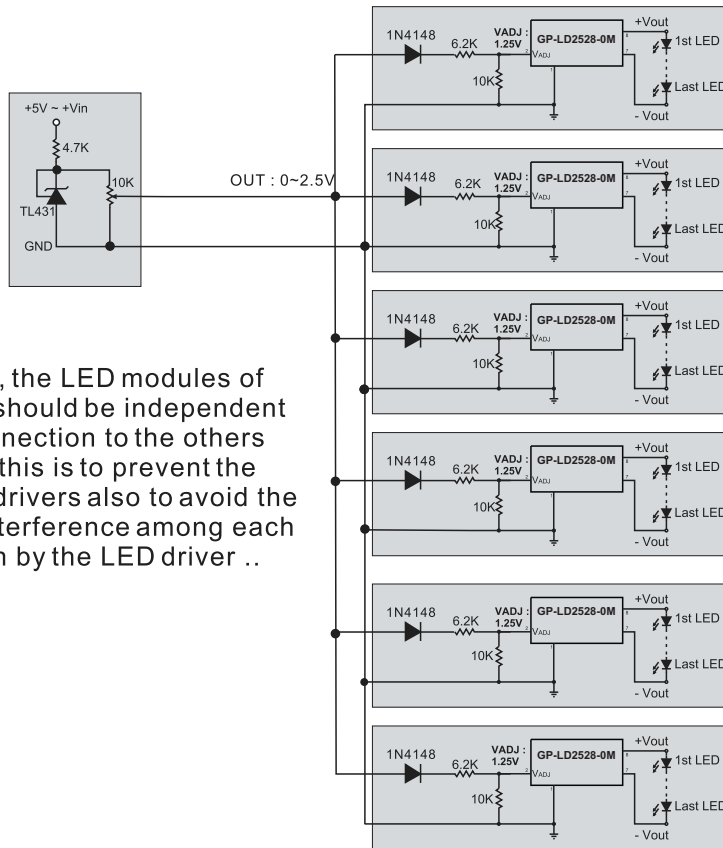



Output Current Adjustment By PWM Control (Flash)



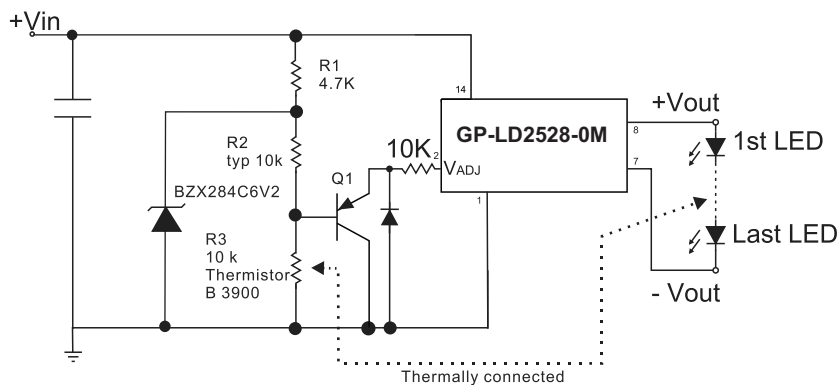
Typical application

Output Current Adjustment By External DC Control Voltage



 In this application, the LED modules of each LED drivers should be independent from electrical connection to the others and input power - this is to prevent the damaging to LED drivers also to avoid the un-necessary interference among each LED module driven by the LED driver ..

Thermal feedback circuit



The selection of components for the thermal feedback circuit is not only dependent on the choice of R2 and R3, but also on the amount of heat sink area required to extract heat from the LEDs. To maximize the light output at high ambient operating temperature conditions, the LEDs must have a sufficient thermal extraction path, otherwise the thermal control circuit will effect current drive reduction in non-optimal conditions. The thermal control threshold point is set by adjusting R2. For this design, three values (33k, 22k and 10k) were evaluated. These values were chosen to give break points at approximately 25° C, 40° C and 60° C. Note that the light output will not continually dim to zero - the thermal control is applying DC control to the ADJ pin and therefore has a dimming ratio from maximum current of approximately 5:1. Once the reduced DC level goes below the shutdown threshold of around 200mV, the LED drive current will fall to zero and the LEDs will be extinguished. The slope of the current reduction is determined by the beta value of the thermistor. The larger the beta value, the sharper will be the resultant current control response. The slope of the current reduction is also affected by Q1's base emitter voltage (VBE) variation with temperature.